

Design and Analysis of Optimal Controllers for Grid Connected Inverters for Photovoltaic Applications

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Abstract

The focus of this research article is to model and analyze optimal controllers for a two level, pulse width modulated, grid connected inverter using Matlab. The Proportional Resonant controller and Linear Quadratic Regulator are being investigated. The controllers are designed such that their performance is satisfactory. The simulation results are presented to illustrate the performance of the designed controllers under different grid conditions.

Keywords: Grid Connected Inverter, Current Control, Linear Quadratic Regulator, Proportional Resonant Controller, Photovoltaic Systems

1. Introduction

The power quality, importance increases in Renewable Energy Systems (RES) especially PhotoVoltaic (PV) systems. The increase in the number of the Photovoltaic systems connected to the grid has increased the importance for the implementation of a unified standard for these installations. In this regard, the standards followed are IEEE 1547 and IEEE929 along with IEC 612727. According to these standards, the overall allowable limit of the Total Harmonic Distortion (THD) is 5%. The odd harmonics from 3rd to 9th should be under 4% each and allowable limit of the odd harmonics from 11th to 15th must be 2%.

These standards present a challenge to the design engineers to develop current control systems that is able to not only meet these requirements, but is also capable of rejection of the grid variation in order to ensure a reliable operation of the installed PV system according to the prescribed standards of IEEE.

Figure 1. shows a sinusoidal waveform with the fundamental frequency of 50 Hz and a distorted waveform in the presence of 3rd, 5th and 7th harmonics. As seen from the figure that the presence of harmonics can result in a reduction of the power quality. The conventional controllers fail to mitigate the harmonics and thus there arises a need to adopt advance current controllers that can effectively remove the harmonics[1-16].

The main objective of this article is to design a grid connected PVS and investigate the performance of Proportional Resonant (PR) and Linear Quadratic Regulator (LQR) by simulating the systems in *Matlab*. The performance of CCs is analyzed under high distortions using the axioms of control theory.

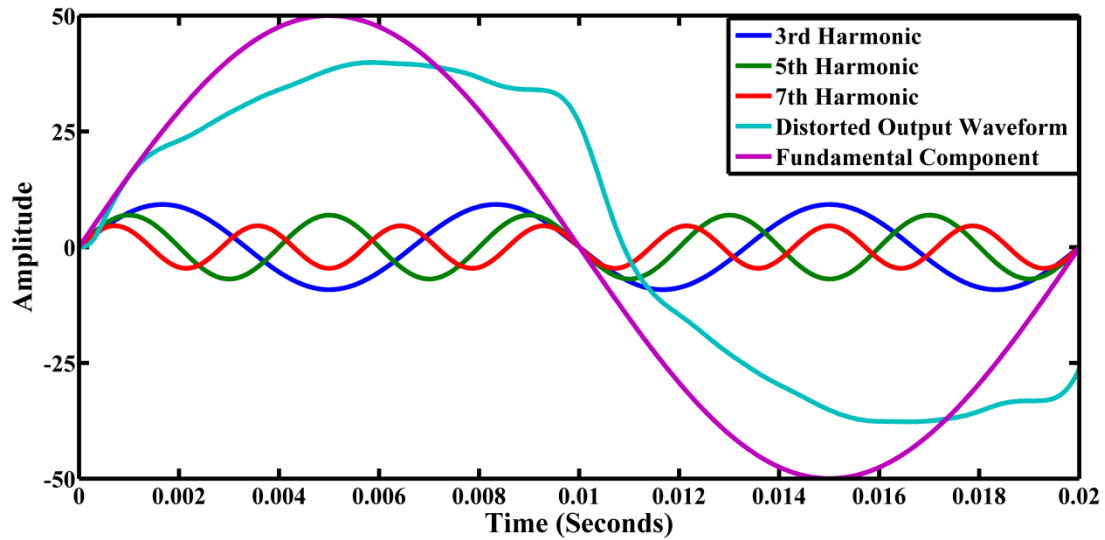


Figure 1. Current Waveform Under Harmonics

2. Mathematical Modelling of the Grid Connected System

The control system forms an integral part of the PVS[1][2][3][4]. The control of the renewable energy system has two main parts. The grid side control and the input power control[6]. The following analysis of the single phase linear model of the converter is given by assuming that the system is balanced and there exists no phase interaction between the phases.

$$V_{input} - V_c = Lg \frac{dI_g}{dt} \quad (1)$$

$$I_c = C \frac{dV_c}{dt} \quad (2)$$

$$I_c = I_g - I_u \quad (3)$$

$$V_c - V_u = L_2 \frac{dI_u}{dt} \quad (4)$$

The expression for the output current can be derived by using the above equations as shown below;

Putting (2) in (3)

$$C \frac{dV_c}{dt} = I_g - I_u$$

$$I_g = \frac{dV_c}{dt} + I_u \quad (5)$$

Differentiating (5)

$$\frac{dI_g}{dt} = \frac{d}{dt} \left[C \frac{dV_c}{dt} \right] + \frac{dI_u}{dt} \quad (6)$$

Putting (6) in (1)

$$V_{input} - V_c = Lg \left[\frac{d}{dt} C \frac{dV_c}{dt} + \frac{dI_u}{dt} \right]$$

$$V_{input} - V_c = LgC \frac{d^2V_c}{dt^2} + Lg \frac{dI_u}{dt}$$

$$V_{input} = LgC \frac{d^2V_c}{dt^2} + Lg \frac{dI_u}{dt} + V_c \quad (7)$$

Substitute the value of $V_c = Lu \frac{dI_u}{dt} + V_{utility}$ in (7)

$$V_{input} = LgC \frac{d^2}{dt^2} (Lu \frac{dI_u}{dt} + V_{utility}) + Lg \frac{dI_u}{dt} + V_c$$

$$V_{input} = Lg \left[\frac{d^2C}{dt^2} (Lu \frac{dI_u}{dt} + V_{utility}) + \frac{dI_u}{dt} \right] + V_c$$

$$V_{input} = Lg \left[\frac{d^2C}{dt^2} (Lu \frac{dI_u}{dt} + V_{utility}) + \frac{dI_u}{dt} \right] + Lu \frac{dI_u}{dt} + V_u \quad (8)$$

$$V_{input} = Lg \left[\frac{d^2}{dt^2} CV_{utility} + CLu \frac{d^3I_u}{dt^3} + \frac{dI_u}{dt} \right] + Lu \frac{dI_u}{dt} + V_{utility}$$

Applying Laplace Transform

$$V_{input} = Lg \left[s^2 CV_{utility} + LuCs^3 I_u + sI_u \right] + LusI_u + V_{utility}$$

$$V_{input} = Lgs^2 CV_{utility} + LgL uCs^3 I_u + LgsI_u + LusI_u + V_{utility}$$

$$V_{input} = Lgs^2 CV_{utility} + (LgL uCs^3 + Lgs + Lus)I_u + V_{utility}$$

$$V_{input} = V_{utility}(Lgs^2C + 1) + I_u(LgL uCs^3 + Lgs + Lus)$$

$$I_u(L_g L_u C_s^3 + L_g s + L_u s) = V_{input} - V_{utility}(L_g s^2 C + 1)$$

$$I_u(L_g L_u C_s^3 + (L_g + L_u)s) = V_{input} - V_{utility}(L_g s^2 C + 1)$$

$$I_u = \frac{1}{(L_g L_u C_s^3 + (L_g + L_u)s)} V_{input} - \frac{(L_g s^2 C + 1)}{(L_g L_u C_s^3 + (L_g + L_u)s)} V_{utility}$$

If the gain for the PWM block is assumed to be unity then the relationship between the output current and reference input current is given by the following transfer function;

$$I_u = \frac{G_z(s)}{1+G_z(s)} I_{ref} - \frac{G_{plant}(s)}{1+G_z(s)} V_{utility} \quad (9)$$

$$G_z(s) = G_{plant}(s) * G_{controller}(s) \quad (10)$$

To make the system more realizable and closer to real time scenarios a disturbance function can be introduced which is given by the following equation as

$$\text{Disturbance}(s) = V_{utility}(L_g C_s^2 + K_c C_s + 0.5) \quad (11)$$

The system values that are used for the Matlab simulation are given in the Table 1.

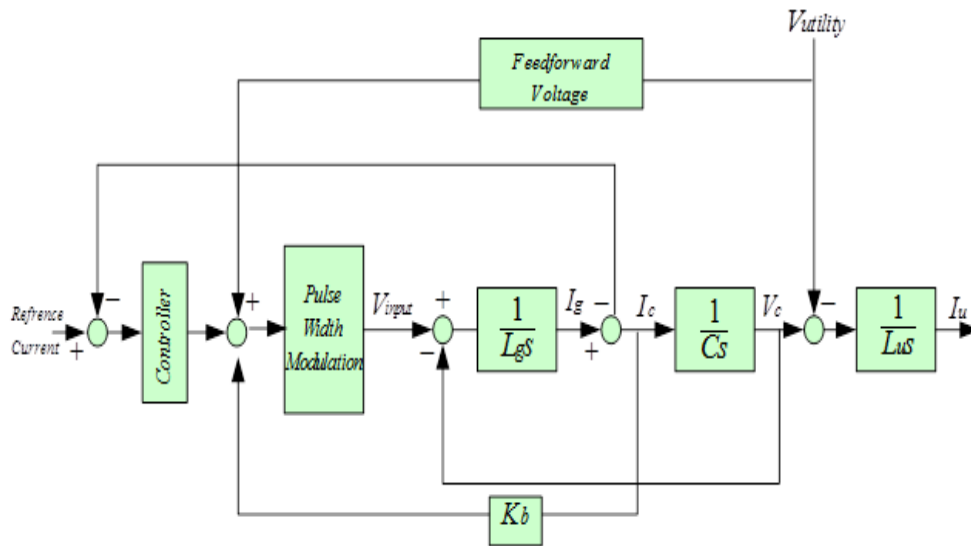


Figure 2. Control Structure of the Grid Connected Inverter

Table 1. Component Values of the Grid Connected System

Sr. No.	Electrical Parameter	Symbol	Value
1	Peak Voltage of the Grid	V_{peak}	230 V (rms)
2	DC link voltage	V_{dc}	800 Vdc
3	Grid inductor	L_g	350 μ H
4	Utility inductor	L_u	50 μ H
5	Capacitor	C_s	22.5 μ F
6	Switching frequency	f_s	10 kHz
7	Grid Frequency	f_g	50Hz
8	Output Current	I_{out}	50A

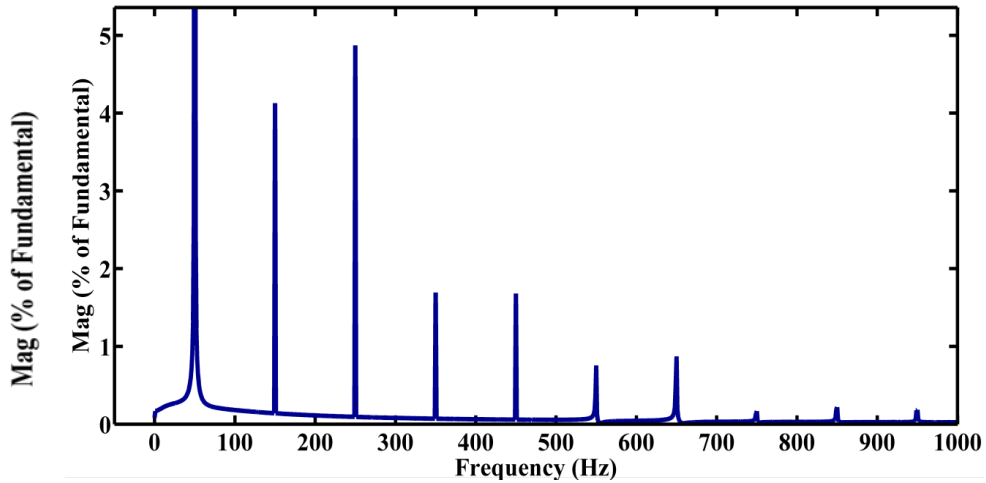


Figure 3. Frequency Spectrum for PI Controller having THD=8.05%

3. Matlab Simulation Results

3.1. Frequency Response of PI Controller Under Harmonic Distortions

Figure 3 shows the Fast Fourier Transform (FFT) analysis of the PI controller when the a THD of 2.74% was introduced in the utility voltage. The THD values at the output was found to be 8.05% . Though it was able to reduce the THD values, but it failed the ANSI-IEEE recommended THD values which is 5 % . The odd harmonic at 250 Hz was the highest harmonic that the PI controller failed to suppress.

3.2. Frequency Response of Optimal Controller Under Harmonic Distortions

A novel Proportional Resonant (PR) controller was designed using Harmonic Compensators (HC) for higher order harmonics. The frequency spectrum analysis was obtained as shown in Figure.

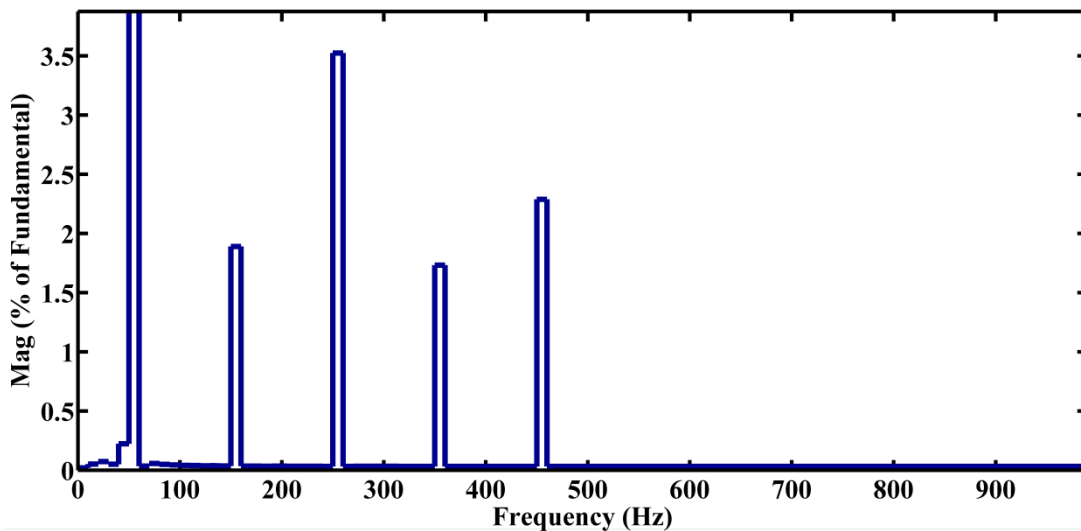


Figure 4. Frequency Spectrum for Proportional Resonant Controller having THD 4.96%

As shown in the Figure that the PR controller was able to achieve the prescribed limits of ANSI-IEEE of 5% of THD. The highest harmonic was observed at 250 Hz (5th Harmonic) with the magnitude of 3.4% of the fundamental frequency. By using the PR controller THD is 0.4 % less than the prescribed limits of recommended THD values. As it was stated earlier, the increase in the performance of the controller was achieved at the cost of the increased complexity at the same time.

3.3. Frequency Spectrum Analysis of Linear Quadratic Regulator (LQR)

The frequency spectrum analysis of the conventional controllers showed that they were unable to meet the operating requirements. In order to reduce the THD to the acceptable limits a Linear Quadratic Regulator (LQR) was designed. The state space model was achieved by using the state space block of Matlab. The values of the gain are given by

$$Q=[0.5 \ 0 \ 0; \ 0 \ 135 \ 0; \ 0 \ 0 \ 10000000]$$
$$R=0.3$$

Figure 5. shows the frequency spectrum of a LQR. The LQR though successively reduced the THD to 4.51%. It is important to note that adjusting the gains of a LQR controller is an iterative process and there is no hard fast rule that states how they should be adjusted. After a trial and error process the LQR controller was being adjusted so that it can reduce the THD to 4.51% as compared to 8.05% of a PI controller. A further adjustment in the gains may have reduced the THD limits further, but it results in the severe performance degradation of the LQR. Therefore, the gain of an LQR was being kept such that it achieves a THD of 4.51%.

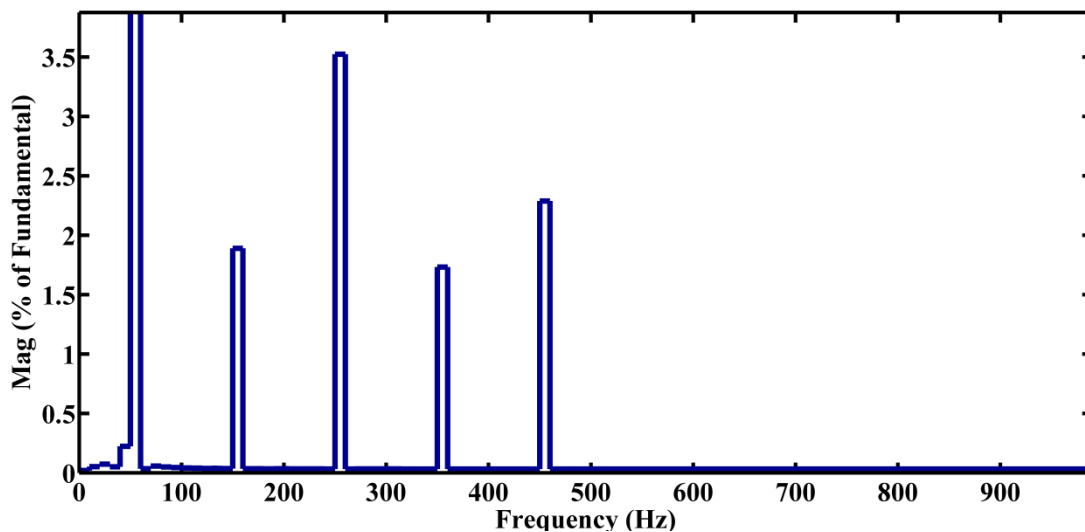


Figure 5. Frequency Spectrum for LQR having THD 4.51%

4. Conclusion

This paper presents an analysis of the optimal controllers for Utility connected PV systems. The results show that the allowed IEEE THD values are 5% and the only controllers that are able to keep the THD values under this limit are optimal controllers PR and LQR. The PI controller is easier to implement, but they are prone to inherent limitations due to which they fail to perform according to the standards of ANSI-IEEE under distortions. In the future, further investigation of both controllers on Real Time Digital Simulator (RTDS) would be done to verify the results obtained from Matlab.

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